A 1280x960 3.75um pixel CMOS imager with Triple Exposure HDR

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A triple exposure 20-bit output high dynamic range (HDR) CMOS imager with sub 2e- rms noise floor is presented. The array consists of 1280x960 pinned photodiode pixels with 3.75um pitch and programmable (high or low) conversion gain (Fig. 1). A two-way row sharing architecture is utilized after weighing fill factor (i.e. QE and crosstalk) against conversion gain (reduced with sharing). The dual conversion gain (CG) feature allows CG adjustment globally across all pixels to match the overall light level in the scene. This is controlled by turning the DCG gate ON or OFF to connect or disconnect the in-pixel capacitor C to the floating diffusion (FD) node. CG is equal to 147uV/e- in High CG (HCG) mode and 37uV/e- in Low CG (LCG) mode. The timing diagram in Fig.2 illustrates the pixel readout process in LCG mode. The DCG gate is turned on when ROW is active, as well as during the extra pixel reset operation. The latter is done to avoid long duty cycle high voltage stress on node A when the sensor is running in LCG mode.

The 'triple exposure' HDR method yields two knee-points and is essentially an extension of the HDR scheme presented by Yadid-Pecht et al (JPL) at this Workshop in 1997. It benefits from inherent signal-to-noise advantages compared to other HDR schemes (e.g. lateral overflow), namely (i) no added pixel circuitry, (ii) no reduction in full-well capacity and (iii) supports regular CDS readout. This comes at a cost of 3x faster readout electronics and on-chip line buffer memory, both of which increases die size and power.

In this device we have implemented 42+3 rows of line memory based on a target to keep SNR above 20dB (in LCG mode) at the knee-points where SNR is at its lowest. The term 'triple exposure' signifies three captures per frame, each with different integration time (T1, T2 and T3). This is achieved by sampling and digitizing three pixel rows per row-time (nominal 22.2usec) as illustrated in (Fig.3). Two of the rows (T1 and T2) are stored in memory for later use and the third (T3) is combined with previously sampled T1 and T2 data to generate the output value. T1 is the longest integration time. Its' maximum value is equal to 1/(Tframe-T2-T3) which is the case illustrated in Fig.3. The T3 row represents the shortest integration time, and it is also the output row. The other two

are 'pushed' onto their corresponding line memory stacks (one for T1 and one for T2). At the opposite end of these stacks one finds the T1 and T2 pixel values corresponding to the output (T3) row. By combining these three values (T1, T2, and T3) one obtains a 20bit linear response pixel value suitable for post processing in the ISP (color processing, tone mapping, etc.).

Fig.4 shows the chip block diagram. One-time-programmablememory (OTPM) is used to store trimming values and pixel defect correction table. The auto-exposure module sets integration time and exposure ratios depending on the scene dynamics. After linearization to 20bits in the signal processing data path, output values can be compressed to 14bits or 12bits (parallel or serial SLVS) in the output data path. A more detailed view of the analog readout architecture is given in Fig.5. Triple exposure HDR imposes three times faster row sampling and A/D conversion compared to a regular linear sensor. Both serial readout (one ADC) and group parallel readout were considered, but a column-parallel architecture was chosen primarily due to lower power consumption. Every column-pair has its own 2:1 multiplexer (MUX), a 20dB programmable gain amplifier and a 12-bit successive approximation (SA) ADC. Even and odd columns are sampled at the top and bottom depending on the row number being odd or even. This gives better layout efficiency (twice the pixel pitch per column layout) and reduced green channel offsets (since neighboring Gr and Gb pixels are read out through the same column mux+amp+ADC). The SA-ADCs are capable of running up to 750kpix/sec. The Correlated Double-Sampling (CDS) and A/D conversion process for one row lasts approximately 5.5us.

The chip floorplan (Fig.6) illustrates the positioning of the various modules. Both serial (1-4 lane SLVS at 450Mbps) and 12-bit parallel output is supported. In the latter case a 20-to-12 bit companding scheme with its knee-points close to the actual T1/T2/T3 knee-points is utilized to minimize compression losses.

At present time, first silicon has emerged from the fab and available measurement data confirms projected performance. Fig. 7 illustrates HDR 20bit output mode signal and noise performance. Measured dynamic range is >110dB. Fig. 8 illustrates signal-to-noise ratio as a function of light level. It confirms that SNR keeps well above 20dB at the kneepoints. Fig. 9 shows SNR performance in linear mode. By selecting LCG or HCG upon light level a dynamic range of >72dB is achieved. Low readout noise <2e- enables image capture with Navitar (F#2.8) lens below 0.05lux at 5 frames per second. Bare die quantum efficiency (QE) is shown in Fig. 10. Maximum QE reaches 60% for green. A captured HDR scene is shown in Fig.11, and some basic performance parameters are listed in Table 1.

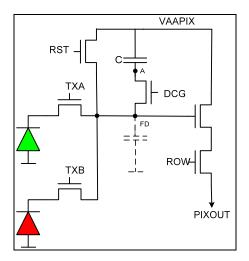


Figure 1

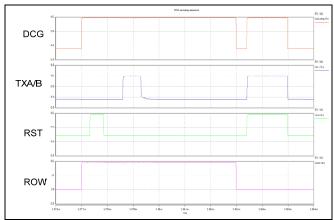


Figure 2

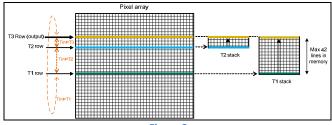


Figure 3

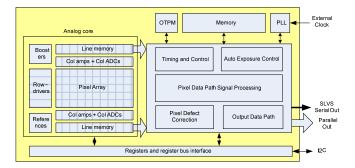


Figure 4

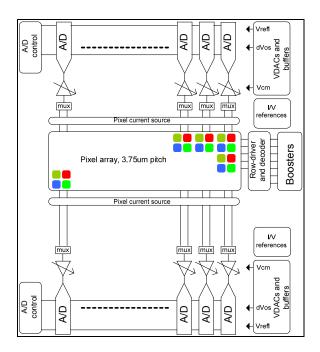


Figure 5

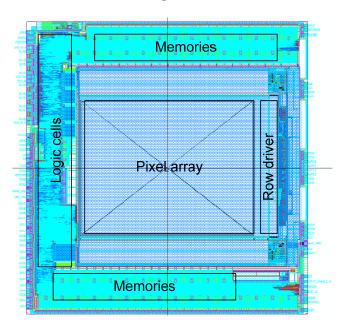
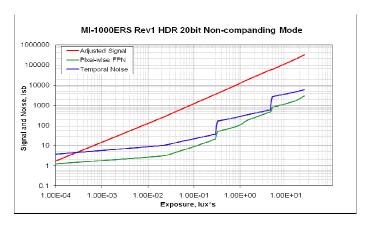


Figure 6

Parameter	Value	Comments
Power consumption	340mW	720p60 mode, Vaa=2.8, Vdd=1.8
High CG	147uV/e-	Measured on 1st silicon
Low CG	37uV/e-	Measured on 1st silicon
Source follower gain	0.92	Measured on 1st silicon
Readnoise @ Low CG	< 7e- rms	10x gain, 74.25Mpix/sec
Readnoise @ High CG	< 2e- rms	10x gain, 74.25Mpix/sec
Dynamic range @1x gain	112dB	CGL, T1/T2=T2/T3=16/1
Dynamic range @8x gain	106dB	CGH, T1/T2=T2/T3=16/1
T1/T2=T2/T3 ratios	4x, 8x, 16x, 32x	Programmable
Output data	12b, 14b, 20b	Parallel or serial (SLVS)
Chip die size	7.9mmx7.9mm	

Table 1



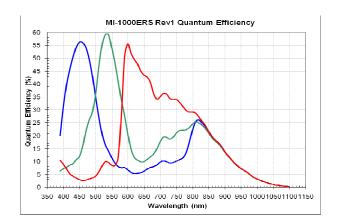
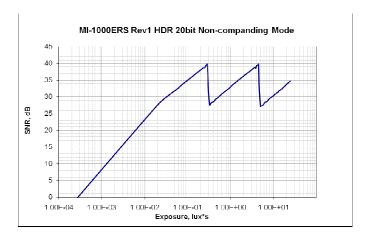


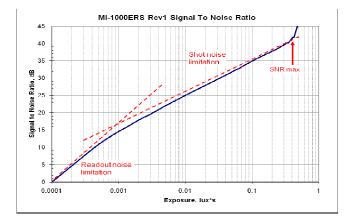
Figure 7 Figure 10



MI-1000ERS Rev1 Fixed Pattern Noise vs. Signal 3 PixelWise FPN Row FPN 2.5 -Column FPN 2 FPN. % 1.5 0.5 0 0 10 20 30 40 50 60 70 80 90 100 Signal, %

Figure 8

Figure 11



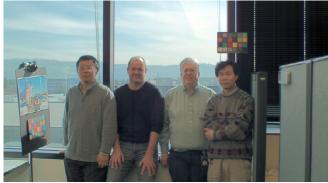


Figure 9

Figure 12